

Static Error Free Tracking Control Based on Linear Quadratic Regulator of 3-DOF Parallel Manipulator with Redundant Actuation

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Abstract: Parallel manipulators have advantages such as high stiffness, low moving inertia, high payload capability, and high accuracy. In this paper, the servo drive system of valve controlled asymmetrical hydraulic cylinder is designed for 4-SPS/S parallel manipulator with redundant actuator. A unified nonlinear model suitable for forward and inverse motion is deduced for the valve controlled asymmetrical cylinder system, and simplified from point of engineering application. The global linearized model is established by nonlinear state feedback transformation. The tracking controller with zero steady-state errors based on linear quadratic regulator is designed for the global linearized model. Simulation results show that tracking performance of the proposed controller is good.

Keywords: Parallel manipulator, static error free tracking control, linear quadratic regulator, redundant actuation.

1. INTRODUCTION

Since Stewart's proposal to employ a parallel manipulator as a flight simulator in 1965 [1], over the last three decades, parallel manipulators are widely used in the entertainment and machine tool industries, especially in motion simulators [2]. They have many distinct advantages compared with the serial manipulators, such as the small accumulated error, fast response speed, high system stiffness, and a higher force-to-weight ratio [3]. The control of parallel manipulator has attracted much attention over the last decade. A number of studies on the control of parallel manipulator have been published, such as Stefano and Lorenzo proposed a force/position control strategies to deal with tasks involving interaction with the environment [4]. Whitey thoroughly analyzed the control and stability issues of the closed-loop force control of a one-link manipulator [5]. Lu and James proposed adaptive synchronized control for a planar parallel manipulator [6]. Djaferis proposed a robust hybrid control algorithm which achieves result by choosing controller's gains that minimize a performance index at a given range of modelling errors [7]. Hag, Young and Kyo proposed a robust nonlinear controller equipped with a friction estimator for a 6 degree of freedom parallel manipulator in the task space coordinates [8], and so on.

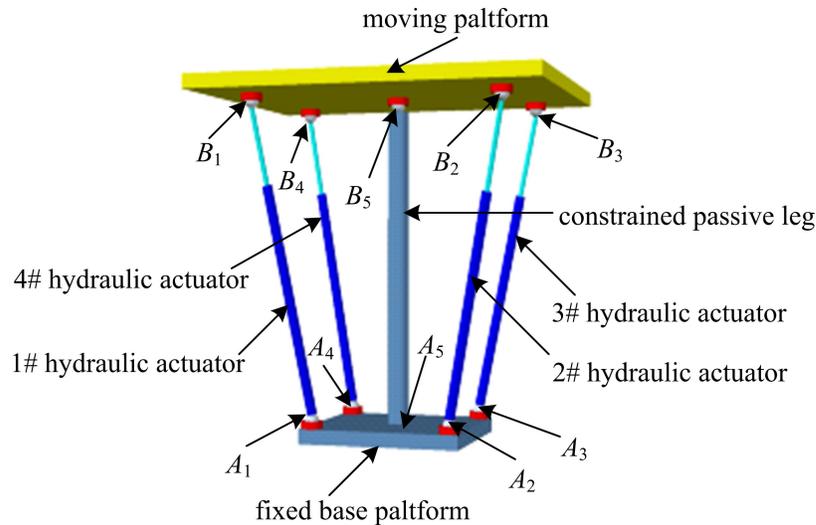
However, accurate trajectory tracking control for parallel manipulators is still a key system requirement, as these devices must often follow prescribed motion [9, 10]. Tracking control of parallel manipulators has been approached using

both linear and nonlinear control laws [11, 12]. High performance control strategies can significantly improve the tracking performance [3, 12]. The dynamics of hydraulic systems are highly nonlinear due to the phenomena such as fluid compressibility, nonlinear servo valve flow-pressure characteristics, and dead band due to the internal leakage and hysteresis [13-15]. However, compared with electrical actuators, hydraulic actuators have many distinct advantages such as the ability to produce large forces at high speeds, the high durability and stiffness, and rapid response [13, 15].

Therefore, this paper will address the tracking control of hydraulic parallel manipulator. The rest of this paper is organized as follows. In Section 2, we present the structure of 4-SPS/S parallel manipulator with redundant actuator and hydraulic drive system. Then a unified nonlinear model suitable for forward and inverse motion is deduced for the valve controlled asymmetrical cylinder system in Section 3. Thereafter, the proposed method for static error free tracking control based on linear quadratic regulator (LQR) is presented in Section 4. In Section 5, the trajectory tracking simulations are carried out. In the end, a short conclusion is drawn in Section 6.

2. SYSTEM OVERVIEW AND MODELLING

The three degree-of-freedom (DOF) hydraulic parallel manipulator includes a moving platform, a fixed base platform, four spherical joint-prismatic joint-spherical joint (SPS) type active legs with the linear hydraulic actuators, and one spherical joint type constrained passive leg, see Fig. (1). Here, the constraint passive leg is perpendicular to the fixed base platform, and connects the moving platform with the fixed platform by a flange plate on the fixed base platform at A_5 and a spherical joint on the moving platform at



$A_1, A_2, A_3, A_4, B_1, B_2, B_3, B_4$ and B_5 are spherical joints
 A_5 is flange plate

Fig. (1). Simplified 3-dof hydraulic parallel manipulator model.

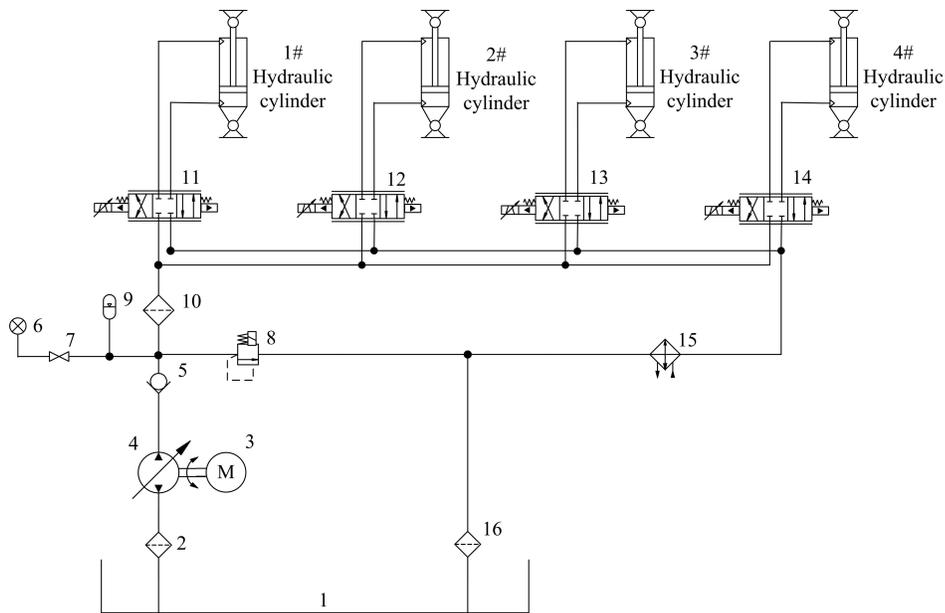


Fig. (2). Diagram of hydraulic drive system.

B_5 . Each of the SPS-type active legs connects the moving platform to the fixed platform by a spherical joint at B_i ($i=1, 2, 3, 4$), an active leg with a prismatic joint, and a spherical joint at A_i ($i=1, 2, 3, 4$), respectively. Quadrilateral $B_1B_2B_3B_4$ is a rectangle with the side $B_1B_2=l_1, B_2B_3=l_2$, and a center point B_5 . Quadrilateral $A_1A_2A_3A_4$ is also a rectangle with the side $A_1A_2=l_3, A_2A_3=l_4$, and a center point A_5 . The length of constraint passive leg is h .

Fig. (2) shows the hydraulic drive system schematic diagram of 4-SPS/S three degree-of-freedom redundant actuated parallel, where element 1 is hydraulic oil tank, element 2 and 16 are crude oil filters, element 3 is motor, element 4 is vari-

able displacement pump, element 5 is check valve, element 6 is pressure gauge, element 7 is pipe joint of the pressure gauge, element 8 is electronic control overflow valve, element 9 is accumulator, element 10 is fine oil filter, element 11, 12, 13 and 14 are electro-hydraulic servo valves, element 15 is cooler.

3. MATHEMATICAL MODELLING OF HYDRAULIC DRIVE SYSTEM

Fig. (3) shows the diagram of hydraulic system with the asymmetric valve controlling asymmetric cylinder, where y

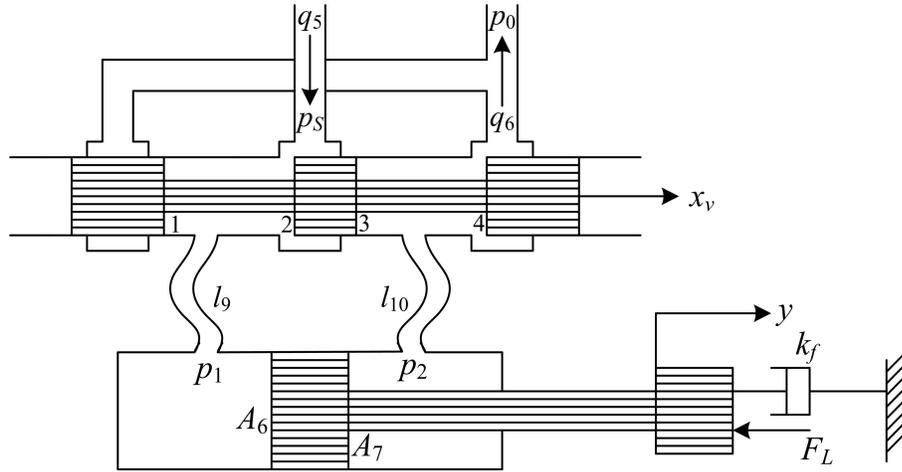


Fig. (3). Diagram of hydraulic system with valve controlled asymmetric cylinder.

is the piston rod displacement, \dot{y} is the piston rod velocity, \ddot{y} is the piston rod acceleration. M_a is the total mass of the piston rod and load, k_f is the viscosity damping coefficient, F_L is the load force, x_v is spool displacement. The oil inlet and return throttle window area-grads of servo valve are represented by w_1 , w_2 , w_3 , and w_4 , where w_1 and w_2 are area-grads of throttle window 1 and throttle window 2, respectively, connecting the servo valve and the hydraulic cylinder no-rod chamber, w_3 and w_4 are area-grads of throttle window 3 and throttle window 4 connecting the servo valve and the hydraulic cylinder rod chamber. p_s and p_0 are supply pressure and spill pressure of system. p_1 , p_2 , q_5 and q_6 are pressure and flow of rod chamber and no-rod chamber. A_6 , A_7 , V_{01} , V_{02} , V_1 and V_2 are effective area, initial volume and volume of rod chamber and no-rod chamber. C_i and C_e are internal leakage coefficient and external leakage coefficient of hydraulic cylinder. C_d is flow coefficient. ρ is density of hydraulic oil. E_5 is bulk modulus of hydraulic oil. l_9 and l_{10} are the lengths of flexible pipe connecting the rod chamber with no-rod chamber. D_1 , h_3 and E_6 are diameter, wall thickness elastic coefficient of flexible pipe. Directional variables are positive direction shown in the Fig. (3).

Assume that the supply pressure p_s is stable and the spill pressure $p_0=0$ in the work process. The oil inlet and return throttle window area-grads of servo valve meet the follow conditions: $w_1=w_2$, $w_3=w_4$ and $\frac{w_3}{w_1} = \frac{w_4}{w_2} = n_0$, where, $0 < n_0 \leq 1$. The effective area of rod chamber and no-rod chamber meet the following condition: $\frac{A_7}{A_6} = m_0$, $0 < m_0 \leq 1$. The $\text{sgn}(\partial)$ is a sign function, $\partial > 0$, $\text{sgn}(\partial) = 1$; $\partial < 0$, $\text{sgn}(\partial) = -1$.

The pressure-flow equation of slide valve is:

$$q_5 = C_d w_1 x_v \sqrt{\frac{1 + \text{sgn}(x_v)}{\rho} (p_s - p_1) + \frac{1 - \text{sgn}(x_v)}{\rho} p_1}, \quad (1)$$

$$q_6 = C_d w_3 x_v \sqrt{\frac{1 - \text{sgn}(x_v)}{\rho} (p_s - p_2) + \frac{1 + \text{sgn}(x_v)}{\rho} p_2}. \quad (2)$$

The flow continuity equation of hydraulic cylinder is:

$$q_5 = A_6 \frac{dy}{dt} + C_i (p_1 - p_2) + C_e p_1 + \xi_1 \frac{dp_1}{dt}, \quad (3)$$

$$q_6 = A_7 \frac{dy}{dt} + C_i (p_1 - p_2) - C_e p_2 - \xi_2 \frac{dp_2}{dt}, \quad (4)$$

Where,

$$\xi_1 = \frac{V_{01} + A_6 y}{E_5} + \frac{l_9 \pi D_1^2}{4} \left(\frac{1}{E_5} + \frac{D_1}{h_3 E_6} \right),$$

$$\xi_2 = \frac{V_{02} - A_7 y}{E_5} + \frac{l_{10} \pi D_1^2}{4} \left(\frac{1}{E_5} + \frac{D_1}{h_3 E_6} \right).$$

The equilibrium equation of hydraulic cylinder is:

$$A_6 p_1 - A_7 p_2 - k_f \frac{dy}{dt} - F_L = M_a \frac{d^2 y}{dt^2}. \quad (5)$$

System state variables are defined as:

$$\mathbf{x} = \begin{bmatrix} x_1 & x_2 & x_3 & x_4 \end{bmatrix}^T = \begin{bmatrix} y & \dot{y} & p_1 & p_2 \end{bmatrix}^T, \quad \text{system}$$

input and output variables are defined as $u=x_v$ and y . Synthesizing equation (1) to equation (5), the state space model of hydraulic drive system is:

$$\begin{cases} \dot{x}_1 = x_2 \\ \dot{x}_2 = \frac{A_6}{M_a} x_3 - \frac{A_7}{M_a} x_4 - \frac{k_f}{M_a} x_2 - \frac{F_L}{M_a} \\ \dot{x}_3 = g_1(\mathbf{x})u - \frac{A_6}{\lambda_1 + \lambda_2 x_1} x_2 - \xi_3 \\ \dot{x}_4 = g_2(\mathbf{x})u + \frac{A_7}{\lambda_3 + m_0 \lambda_2 x_1} x_2 + \xi_4 \end{cases}, \quad (6)$$

Where,

$$g_1(\mathbf{x}) = \frac{C_d w_1}{\lambda_1 + \lambda_2 x_1} \sqrt{\frac{1 + \text{sgn}(u)}{\rho} (p_s - x_3) + \frac{1 - \text{sgn}(u)}{\rho} x_3},$$

$$g_2(\mathbf{x}) = -\frac{C_d w_3}{\lambda_3 + m_0 \lambda_2 x_1} \sqrt{\frac{1 - \text{sgn}(u)}{\rho} (p_s - x_4) + \frac{1 + \text{sgn}(u)}{\rho} x_4}$$

$$u = x_v, \lambda_1 = \frac{V_{01}}{E_5} + \frac{m_0 \lambda_2 x_1}{4} \left(\frac{1}{E_5} + \frac{D_1}{h_3 E_6} \right), \lambda_2 = \frac{A_6}{E_5},$$

$$\lambda_3 = \frac{V_{02}}{E_5} + \frac{l_{10} \pi D_1^2}{4} \left(\frac{1}{E_5} + \frac{D_1}{h_3 E_6} \right),$$

$$\xi_3 = \frac{C_i}{\lambda_1 + \lambda_2 x_1} (x_3 - x_4) - \frac{C_e}{\lambda_1 + \lambda_2 x_1} x_3,$$

$$\xi_4 = \frac{C_i}{\lambda_3 + m_0 \lambda_2 x_1} (x_3 - x_4) - \frac{C_e}{\lambda_3 + m_0 \lambda_2 x_1} x_4.$$

4. HYDRAULIC DRIVE SYSTEM INPUT-OUTPUT FEEDBACK LINEARIZATION

The servo valve is directly installed on the cylinder of hydraulic drive system, so the length of flexible pipe is very short, and it can be ignored in engineering. Furthermore, because of $\frac{V_{01}}{E_5} \gg \frac{A_6}{E_5}$, the nonlinear state space equation (6) is transformed to affine nonlinear system, and its standard form is:

$$\begin{cases} \dot{\mathbf{x}} = \mathbf{f}(\mathbf{x}) + \mathbf{g}(\mathbf{x})u + \mathbf{Q}_1 \\ y = h(\mathbf{x}) \end{cases}, \quad (7)$$

Where,

$$\mathbf{f}(\mathbf{x}) = \begin{bmatrix} x_2 \\ \frac{A_6}{M_a} x_3 - \frac{A_7}{M_a} x_4 - \frac{k_f}{M_a} x_2 \\ -\frac{A_6}{\lambda_1} x_2 - \frac{C_i}{\lambda_1} (x_3 - x_4) - \frac{C_e}{\lambda_1} x_3 \\ \frac{A_7}{\lambda_3} x_2 + \frac{C_i}{\lambda_3} (x_3 - x_4) - \frac{C_e}{\lambda_3} x_4 \end{bmatrix},$$

$$\mathbf{g}(\mathbf{x}) = \begin{bmatrix} 0 & 0 & g_1(\mathbf{x}) & g_2(\mathbf{x}) \end{bmatrix}^T, u = x_v, h(\mathbf{x}) = x_1,$$

$$\mathbf{Q}_1 = \begin{bmatrix} 0 & -\frac{F_L}{M_a} & 0 & 0 \end{bmatrix}^T.$$

By the equation (7) can be obtained:

$$L_{\mathbf{r}} h(\mathbf{x}) = \frac{\partial h}{\partial \mathbf{x}} \mathbf{f}(\mathbf{x}) = x_2, L_{\mathbf{g}} h(\mathbf{x}) = \frac{\partial h}{\partial \mathbf{x}} \mathbf{g}(\mathbf{x}) = 0, L_{\mathbf{g}} L_{\mathbf{r}} h(\mathbf{x}) = 0,$$

$$L_{\mathbf{r}}^2 h(\mathbf{x}) = \frac{A_6}{M_a} x_3 - \frac{A_7}{M_a} x_4 - \frac{k_f}{M_a} x_2,$$

$$L_{\mathbf{g}} L_{\mathbf{r}}^2 h(\mathbf{x}) = \frac{A_6}{M_a} g_1(\mathbf{x}) - \frac{A_7}{M_a} g_2(\mathbf{x}),$$

$$L_{\mathbf{r}}^3 h(\mathbf{x}) = -\frac{k_f}{M_a^2} (A_6 x_3 - A_7 x_4 - B_c x_2) - \xi_5 - \xi_6,$$

$$\xi_5 = \frac{A_6}{M_a \lambda_1} (A_6 x_2 + C_i (x_3 - x_4) + C_e x_3),$$

$$\xi_6 = \frac{A_7}{M_a \lambda_3} (A_7 x_2 + C_i (x_3 - x_4) - C_e x_4).$$

Select the coordinate transformation is:

$$\mathbf{z} = \Phi(\mathbf{x}) = \begin{bmatrix} \phi_1(\mathbf{x}) \\ \phi_2(\mathbf{x}) \\ \phi_3(\mathbf{x}) \\ \phi_4(\mathbf{x}) \end{bmatrix} = \begin{bmatrix} h(\mathbf{x}) \\ L_{\mathbf{r}} h(\mathbf{x}) \\ L_{\mathbf{r}}^2 h(\mathbf{x}) \\ \eta(\mathbf{x}) \end{bmatrix}. \quad (8)$$

According to the reference [16], a function $\eta(\mathbf{x})$ which makes the map $\Phi(\mathbf{x})$ non-singular Jacobian matrix at \mathbf{x}_0 can be found, and $\eta(\mathbf{x})$ can be selected as:

$$L_{\mathbf{g}} \eta(\mathbf{x}) = \frac{\partial \eta}{\partial \mathbf{x}} \mathbf{g}(\mathbf{x}) = 0. \quad (9)$$

So, the equation (9) can be transformed:

$$\frac{\partial \eta}{\partial \mathbf{x}} \mathbf{g}(\mathbf{x}) = \frac{\partial \eta}{\partial x_3} g_1(\mathbf{x}) + \frac{\partial \eta}{\partial x_4} g_2(\mathbf{x}) = 0. \quad (10)$$

The equation (11) is a solution of equation (10):

$$\eta(\mathbf{x}) = \lambda_1 w_3 \sqrt{[1 + \text{sgn}(u)](p_s - x_3) + [1 - \text{sgn}(u)]x_3} + \xi_7, \quad (11)$$

Where,

$$\xi_7 = \lambda_3 w_1 \sqrt{[1 - \text{sgn}(u)](p_s - x_4) + [1 + \text{sgn}(u)]x_4}.$$

After calculation shows Jacobian matrix of $\Phi(\mathbf{x})$ is nonsingular, so the transformation of equation (8) is a global differential homeomorphism, and the original system is transformed as:

$$\begin{cases} \dot{\mathbf{z}} = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} z_1 \\ z_2 \\ z_3 \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} v + \tilde{\mathbf{Q}}_1 \\ y = z_1 \end{cases}, \quad (12)$$

Where,

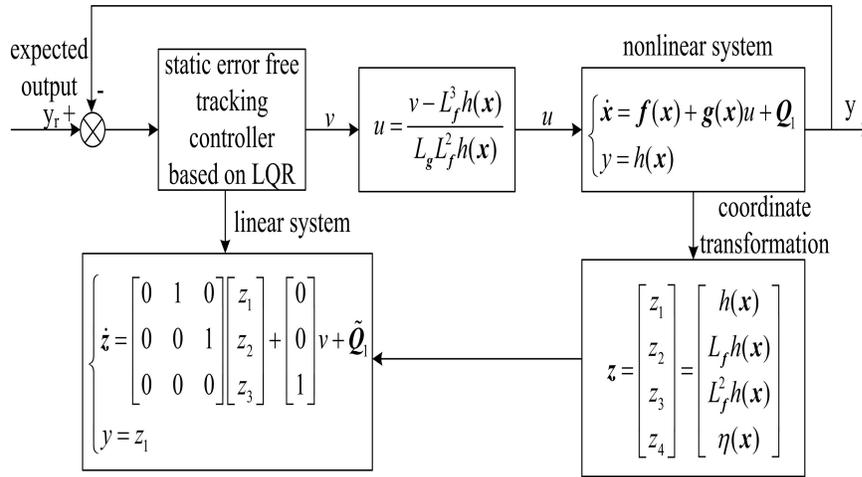


Fig. (4). Scheme of control system.

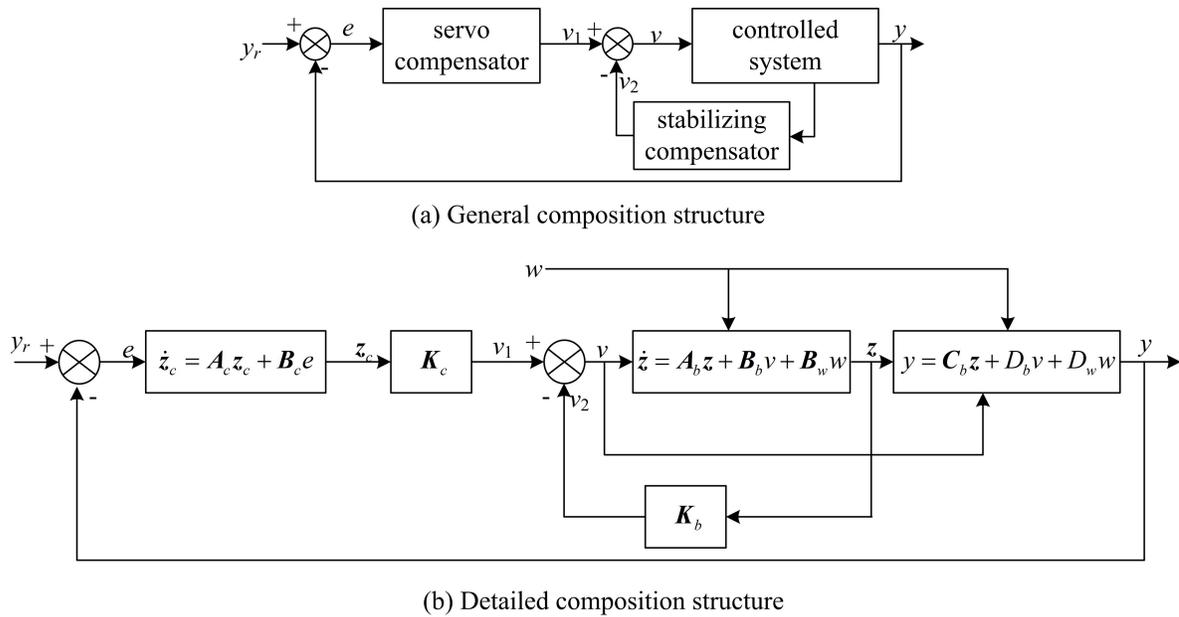


Fig. (5). Static error free tracking control system.

$$\tilde{Q}_1 = \begin{bmatrix} 0 & -\frac{F_L}{M_a} & \frac{k_f F_L}{M_a^2} \end{bmatrix}^T, \quad v = L_f^3 h(x) + L_g L_f^2 h(x) u.$$

According to controllability criterion of linear system, the linear system of equation (12) is completely controllable. The equation (12) represents the external state of system, but the internal state η becomes unobservable because of state feedback. For precise input-output linearization system, the external state should be stable and have a good dynamic quality, and the internal state η should also be made stable as well. According to the reference [16] analysis, the internal state η of system is stable.

The state space model of hydraulic drive system can be transformed into the linear system of equation (12) by

feedback linearization method, so the controller can be designed by linear system method. In this paper, the static error free tracking control based on LQR is used to control hydraulic drive system, and the structure of control system is shown in Fig. (4). The general composition structure of static error free tracking control is shown in Fig. (5a), and the controller consists of servo compensator and stabilizing compensator. The servo compensator is a linear time invariant system, and its function is to control the system to achieve asymptotic tracking and disturbance rejection. The stabilizing compensator is a static state feedback, and its function is to make the control system achieve asymptotical stability. Fig. (5b) is the detailed composition structure of static error free tracking control.

Let the controlled system be completely controlled, and its state space is described as:

$$\begin{cases} \dot{\mathbf{z}} = \mathbf{A}_b \mathbf{z} + \mathbf{B}_b v + \mathbf{B}_w w \\ y = \mathbf{C}_b \mathbf{z} + D_b v + D_w w \end{cases} \quad (13)$$

Where, $n = \dim(\mathbf{z})$, $p = \dim(v)$, $q = \dim(y)$ and w is disturbance signal. The specific composition structure of static error free tracking control is shown in Fig. (5b), and the tracking error is $e(t) = y_r(t) - y(t)$.

The static error free tracking control algorithm based on LQR is as follows:

Step 1. Judge the relationship between the input dimensions and the output dimensions of controlled system, and analyze the controllability of controlled system. Furthermore, the input and output satisfy the following formula: $\dim(v) \geq \dim(y)$.

Step 2. Derive the common instability polynomial $\phi(s)$ of $y_r(t)$ and $w(t)$, i.e. when $t \rightarrow \infty$, derive the frequency domain structure characteristics $\phi_r(s)$ and $\phi_w(s)$ by no zero parts $y_{rs}(t)$ and $w_s(t)$, and calculate the minimal common multiple $\phi(s)$ between $\phi_r(s)$ and $\phi_w(s)$, and $\phi(s) = s^l + \tilde{\alpha}_{l-1}s^{l-1} + \dots + \tilde{\alpha}_1s + \tilde{\alpha}_0$.

Step 3. Judge the rank relationship which can realize static error free tracking control, i.e. each root λ_i of $\phi(s) = 0$ can satisfy the following formula:

$$\text{rank} \begin{bmatrix} \lambda_i \mathbf{I} - \mathbf{A}_b & \mathbf{B}_b \\ -\mathbf{C}_b & \mathbf{D}_b \end{bmatrix} = n + q, \quad i = 1, 2, \dots, l.$$

Step 4. Derive the block coefficient matrix:

$$\Gamma = \begin{bmatrix} 0 & & & & \\ \vdots & & \mathbf{I}_{l-1} & & \\ 0 & & & & \\ -\tilde{\alpha}_0 & -\tilde{\alpha}_1 & \dots & -\tilde{\alpha}_{l-1} & \end{bmatrix}, \quad \beta = \begin{bmatrix} 0 \\ \vdots \\ 0 \\ 1 \end{bmatrix},$$

And derive the common instability model $y_r(t)$ and $w(t)$:

$$\dot{\mathbf{z}}_c = \mathbf{A}_c \mathbf{z}_c + \mathbf{B}_c e,$$

Where, $\mathbf{A}_c = \begin{bmatrix} \Gamma & \\ & 0 \end{bmatrix}$, $\mathbf{B}_c = \begin{bmatrix} \beta & \\ & 0 \\ & & \beta \end{bmatrix}$.

Step 5. Build the state equation of $(n+ql)$ dimensional series system Σ_T :

$$\begin{bmatrix} \dot{\mathbf{z}} \\ \dot{\mathbf{z}}_c \end{bmatrix} = \begin{bmatrix} \mathbf{A}_b & 0 \\ -\mathbf{B}_c \mathbf{C}_b & \mathbf{A}_c \end{bmatrix} \begin{bmatrix} \mathbf{z} \\ \mathbf{z}_c \end{bmatrix} + \begin{bmatrix} \mathbf{B}_b \\ -\mathbf{B}_c \mathbf{D}_b \end{bmatrix} v + \begin{bmatrix} \mathbf{B}_w \\ -\mathbf{B}_c \mathbf{D}_w \end{bmatrix} w + \begin{bmatrix} 0 \\ \mathbf{B}_c \end{bmatrix} y_r$$

Step 6. Establish $p \times (n+ql)$ dimensional matrix \mathbf{K}_T based on LQR method.

Step 7. Matrix \mathbf{K}_T partition is the following formula:

$$\mathbf{K}_T = \begin{bmatrix} -\mathbf{K}_b & \mathbf{K}_c \\ p \times n & p \times ql \end{bmatrix}.$$

Step 8. Derive the stabilizing compensator: $v_2 = \mathbf{K}_b \mathbf{z}$.

Step 9. Derive the servo compensator:

$$\dot{\mathbf{z}}_c = \mathbf{A}_c \mathbf{z}_c + \mathbf{B}_c e, \quad v_1 = \mathbf{K}_c \mathbf{z}_c.$$

The static error free tracking control algorithm based on LQR is applied to 3-DOF parallel manipulator with redundant actuation for trajectory tracking control, and the control strategy is shown in Fig. (6). Each active branch with load force compensation calculation model adopts the static error free tracking control algorithm based on LQR to implement control, where the closed loop control system is composed of the inverse position as desired trajectory of the hydraulic cylinder and the displacement sensor mounted inside the hydraulic cylinder as position feedback signal.

5. SIMULATION RESULTS

The design parameters of the parallel manipulator and hydraulic system are as follows: $l_1=2.4\text{m}$, $l_2=3.6\text{m}$, $l_3=1.8\text{m}$, $l_4=2.0\text{m}$, $h=4.7\text{m}$, $w_1=w_2=3.76 \times 10^{-2}\text{m}$, $w_3=w_4=1.88 \times 10^{-2}\text{m}$, $k_f=850\text{N/(m/s)}$, $A_6=6.362 \times 10^{-3}\text{m}^2$, $A_7=3.244 \times 10^{-3}\text{m}^2$, $V_{01}=1.145 \times 10^{-2}\text{m}^3$, $V_{02}=5.893 \times 10^{-2}\text{m}^3$, $C_i=2.836 \times 10^{-13}\text{m/(s.Pa)}$, $C_e=7.654 \times 10^{-13}\text{m/(s.Pa)}$, $C_d=0.67$, $\rho=875\text{kg/m}^3$, and $E_5=750\text{MPa}$. Since the structure of hydraulic drive system and controller are same, so 1# active branched chain is selected for experiment in order to analyze the effect of the controller. The load force can be calculated by dynamics, it gets the load force $F_L \in [-6000\text{N}, 6000\text{N}]$, so the load force compensation model is selected, $F_L = 6000 \sin(\frac{\pi}{4}t)$.

The control system corresponding to equation (13) is as follow:

$$\mathbf{A}_b = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ 0 & 0 & 0 \end{bmatrix}, \quad \mathbf{B}_b = \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}, \quad \mathbf{B}_w = \begin{bmatrix} 0 \\ -1/M_a \\ k_f/M_a^2 \end{bmatrix}, \quad w = F_L,$$

$$\mathbf{C}_b = \begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix}^T, \quad D_b = D_w = 0.$$

So, the input and output of control system satisfy the following formula: $\dim(v) \geq \dim(y)$, and it is completely controllable. It has two roots $\lambda_1 = -\frac{\pi}{4}j$ and $\lambda_2 = \frac{\pi}{4}j$ for the $\phi(s) = 0$, and the following result can be obtained by calculation:

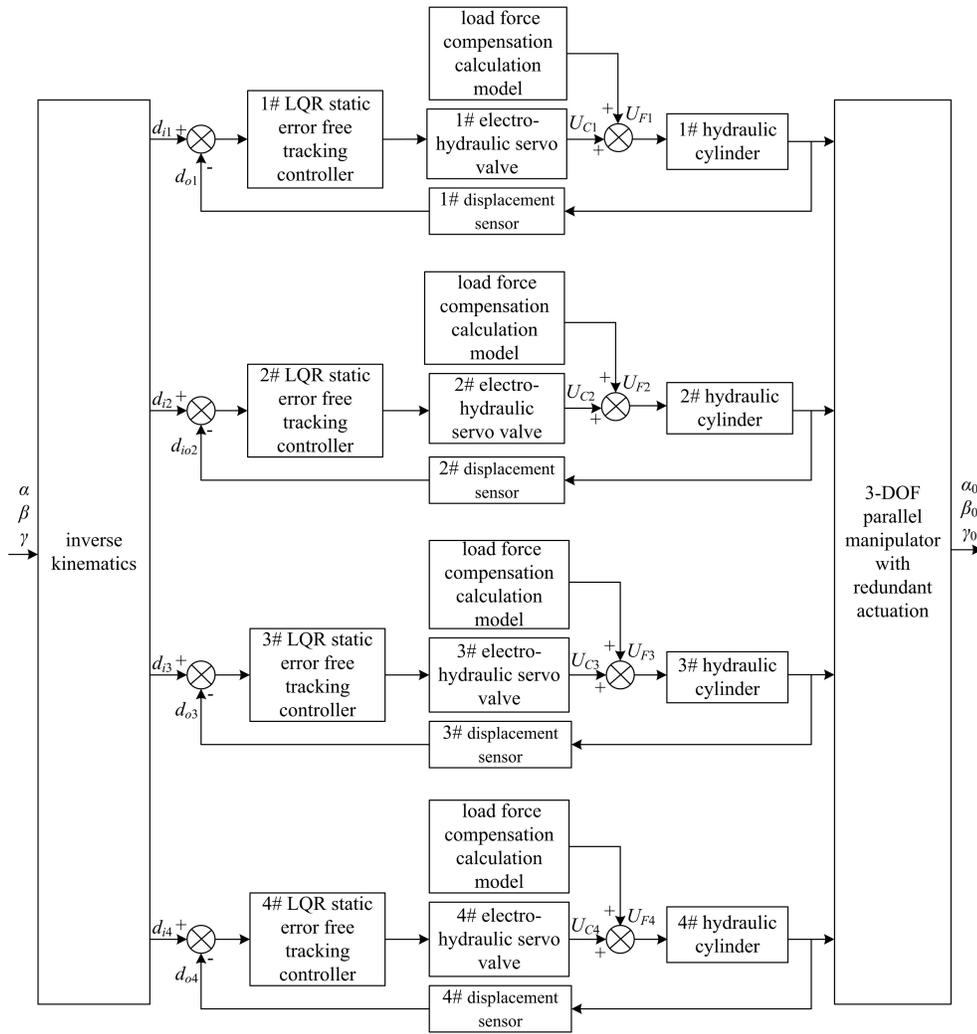


Fig. (6). Control strategy of 3-DOF parallel manipulator with redundant actuation.

$$\text{rank} \begin{bmatrix} \lambda_1 \mathbf{I} - \mathbf{A}_b & \mathbf{B}_b \\ -\mathbf{C}_b & \mathbf{D}_b \end{bmatrix} = \text{rank} \begin{bmatrix} \lambda_2 \mathbf{I} - \mathbf{A}_b & \mathbf{B}_b \\ -\mathbf{C}_b & \mathbf{D}_b \end{bmatrix} = 4 = n + q.$$

Then it can satisfy the rank relationship meeting static error free tracking control.

The common instability model of $y_r(t)$ and $w(t)$ is:

$$\dot{\mathbf{z}}_c = \mathbf{A}_c \mathbf{z}_c + \mathbf{B}_c e = \begin{bmatrix} 0 & 1 \\ -\frac{\pi^2}{16} & 0 \end{bmatrix} \mathbf{z}_c + \begin{bmatrix} 0 \\ 1 \end{bmatrix} e.$$

The state equation of $(n+q)$ dimensional series system Σ_T is:

$$\begin{bmatrix} \dot{\mathbf{z}} \\ \dot{\mathbf{z}}_c \end{bmatrix} = \begin{bmatrix} \mathbf{A}_b & 0 \\ -\mathbf{B}_c \mathbf{C}_b & \mathbf{A}_c \end{bmatrix} \begin{bmatrix} \mathbf{z} \\ \mathbf{z}_c \end{bmatrix} + \begin{bmatrix} \mathbf{B}_b \\ -\mathbf{B}_c \mathbf{D}_b \end{bmatrix} v + \begin{bmatrix} \mathbf{B}_w \\ -\mathbf{B}_c \mathbf{D}_w \end{bmatrix} w + \begin{bmatrix} 0 \\ \mathbf{B}_c \end{bmatrix} y_r \quad (14)$$

The matrix values are substituted into the equation (14), by finishing can be as follow:

$$\begin{bmatrix} \dot{\mathbf{z}} \\ \dot{\mathbf{z}}_c \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 1 \\ -1 & 0 & 0 & -\frac{\pi^2}{16} & 0 \end{bmatrix} \begin{bmatrix} \mathbf{z} \\ \mathbf{z}_c \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 1 \\ 0 \\ 0 \end{bmatrix} \quad (15)$$

$$v + \begin{bmatrix} 0 \\ -1/M_a \\ k_f/M_a^2 \\ 0 \\ 0 \end{bmatrix} F_L + \begin{bmatrix} 0 \\ 0 \\ 0 \\ 0 \\ 1 \end{bmatrix} y_r$$

Easy to determine the system (15) is controllable, and apply LQR method to calculate the whole state feedback gain. It selects the weighting matrix $R_T=[1]$ and $Q_T=\text{diag}([2$

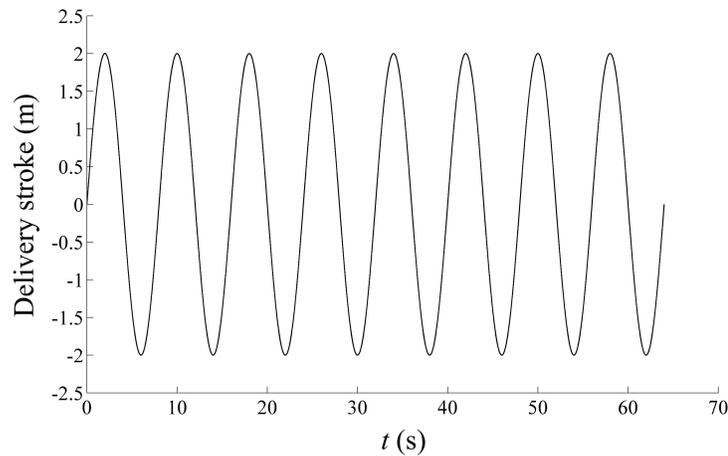


Fig. (7). Stroke tracking curve.

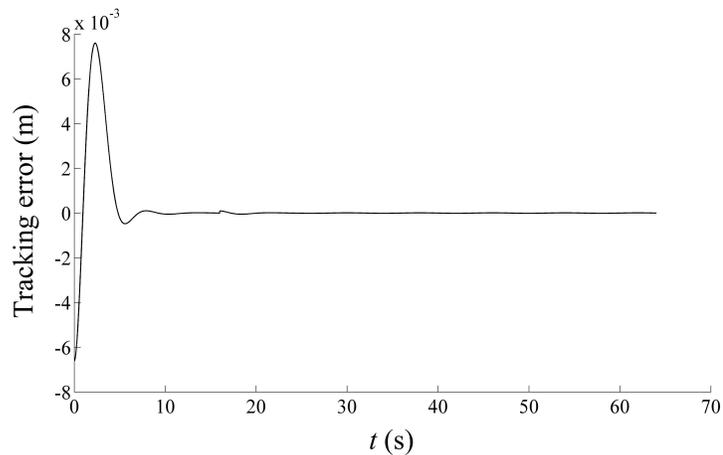


Fig. (8). Tracking error curve.

0 1 1 4]), then the whole state feedback gain can be calculated by calling the function `lqr()` of Matlab, it can obtain the whole state feedback gain $K_1=[13.8003 \ 12.3432 \ 5.8895 \ 4.3163 \ -5.0999]$. Simulation studies are conducted on Matlab-Simulink and 1# active branched input stroke is

$y_r = 2\sin(\frac{\pi}{4}t)$. Fig. (7) shows the results of tracking 1# active branched input stroke and Fig. (8) shows the tracking error.

6. CONCLUSION

Hydraulic servo drive system has been designed for 4SPS/S 3-DOF parallel manipulator with redundant actuation. A unified nonlinear model has been theoretically established for forward and inverse motion of hydraulic cylinder, and the nonlinear model has been simplified from the perspective of engineering applications. On the basis of differential geometry linearization method, the global linear model is achieved by nonlinear state feedback transforming. Thereafter, on the basis of load force compensation, the static error

free tracking controller based on LQR is designed. Simulations demonstrated the effectiveness of the proposed controller, and it can provide a theoretical basis for subsequent practical engineering implementation.

CONFLICT OF INTEREST

The author confirms that this article content has no conflict of interest.

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